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Controlled Rectifier with Improved Dynamic Response for Under Unbalanced Supply and Variable Line and Load Inductances

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ABSTRACT: To simulate and implementation of controlled rectifier with improved dynamic response for under unbalanced supply and variable line and load inductances. With the widespread use of power electronics devices such as rectifier, inverter etc. in power system causes serious problem relating to power quality. One of such problem is generation of current and voltage harmonics causing distortion of load waveform, voltage fluctuation, voltage dip, heating of equipment etc. Also presence of non-linear loads such as UPS, SMPS, speed drives etc. causes the generation of current harmonics in power system. They draw reactive power components of current from the AC mains, hence causing disturbance in supply current waveform. Thus to avoid the consequences of harmonics we have to compensate the harmonic component in power utility system. To simulate controlled Rectifier with open loop system. To simulate controlled Rectifier with closed loop FOPID controller system.

KEYWORDS: Controlled Rectifier, Pulse width modulation (PWM). Total harmonic distortion (THD), Proportional integral (PI), Fractional order Proportional integral derivative (FOPID).

I. INTRODUCTION

Extensive use of power electronic systems gained concern about the power quality. Prominent among power electronics loads are three phase rectifiers. They are widely used in variable speed drives, power supplies systems. Power electronics facilitates conversion and control of electrical power between source and load with the help of electronic switching devices. The conversion and control of voltage, current, frequency, wave shape and power are done by static power converters. Efficiency, power density, power quality, accuracy of control and simple topologies are requirements of power converters. The cost, effectiveness, reliability, robustness, space and weight managements are important in the design and operation of these converters. Among the static power converters, three phase AC-DC converters are used in industrial and nonindustrial applications that require medium and high power AC-DC conversion. AC-DC converters are directly used in the dc drives, battery chargers, electroplating, chemical processes and magnet power supplies. They are extensively used in domestic, commercial and industrial applications. Single and three phase rectifiers are front end components of dc power supplies, electronic ballast, inverters for ac drives and industrial heating and HVDC transmission etc. Problems of AC-DC rectifiers regarding power quality issues are subject of research and investigation. Power quality deterioration rises from the fact that the interface of load with a circuit by a rectifier makes overall circuit a nonlinear one. This is because diodes used in rectifiers are nonlinear in character.



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Nonlinearity causes generation of harmonics that result in poor power quality, voltage distortion, low power factor at input side, slow varying rippled dc output at load end and low efficiency. Harmonics of input current causes overheating of transformers, power cables and motors, inadvertent tripping of relays, incorrect measurement of voltage and current by meters and increased iron losses in transformers. Harmonics also cause motors to heat up and produce pulsating torque. Buildings with large number of computers and data processing equipment use AC-DC conversion for their supply. AC-DC converters draw harmonic current which cause neutral current. Besides power quality problems at user ends and utility, harmonics also cause problems on other users. A method to reduce harmonic pollution caused by the three phase rectifiers is the third harmonic current injection. The technique applies injection of third harmonic currents in the rectifier supply lines in order to modify the input current waveform &to reduce its distortion.

Nowadays step-up converters are used in many applications, especially in renewable energy systems such as those based on PV panels and fuel cells. These converters usually require to step-up low voltages in order to meet the nominal requirements of grid-connected inverters. Currently, there exist several topologies with high volt- age conversion ratios, e.g. those that are based on coupled-inductors. For instance, in [1] the authors introduce Y-source boost DC/DC converter whose high voltage gains are enabled via a tightly coupled three-winding inductor. This converter shows in general interesting design and dynamic properties that are suitable for distributed generation applications. In [2], a high voltage gain multi-coupled inductor based topology is introduced; this topology exhibits automatic current balancing and high-power density, with respect to other coupled-inductor based topologies, since its multi-coupled inductor design per- mits to optimize the use of the core material. Other high voltage gain topologies that are based on coupled inductors can be found in [3]-[10]. Although such devices represent a very plausible solution at achieving high conversion ratios and overall good performance, they usually require non-standard (i.e. noncommercial) magnetic components for their design. Moreover, the required size of the overall converter might be also larger than that of topologies that require only small inductors, due to the need of bulky core materials. As an alternative to these configurations, transformer-less topologies can be constructed with devices off-the-shelf, which is a compelling solution that permits to reduce the design process, size and overall cost of the converter. Such transformerless topologies, can be J.E. Valdez-Resendiz, J.C. Mayo-Maldonado, J. Loranca-Coutio, C.A. Villarreal-Hernandez and G. Escobar-Valderrama are with School of Engineering and Sciences at Tecnologico de Monterrey, Monterrey,

J.C. Rosas-Caro is with Academia de Electrica, Electronica y Control at Universidad Panamericana, Zapopan, Mexico. based on several strategies such as: (i) switched-capacitors as voltage multipliers (see [11]–[16]), with linear voltage gains; (ii) electrical configurations that allow the connection of input parallel- and output series- converter stages to improve performance and achieve large gains (see [17]–[20]); (iii) quadratic gain stages (see [21]–[24]), with important features e.g. high gain with a reduced number of switches. Though, e.g. switched capacitor converters have become a popular solution to achieve high-voltage gains, they also have some disadvantages. For instance they exhibit high current spikes due to due to abrupt parallel connections between capacitors leading to to high conduction losses. Moreover, there exist also charging/discharging losses that reduce the efficiency of such converters (cf. [25]).

In this paper, we focus on providing a new solution for the specific challenging need to develop new power converter topologies for renewable energy applications. In particular for those scenarios that demand high voltage gains, low cost implementations and high efficiency. In particular, we propose a novel step-up DC-DC converter with quadratic gains, whose main advantages are (i) simple single-switch structure, (ii) low voltage rated capacitors compared with traditional quadratic boost converters (QBC), (iii) modularity, several switching stages can be easily stacked in order to increase the output voltage gain. The underlying operation principle of the pro- posed converter is validated with experimental results

II. CONTROLLED RECTIFIER WITH IMPROVED DYNAMIC RESPONSE

2.1 Circuit-Description

The three-phase bridge rectifier suffers from waveform distortion and has high THD at the input side. Reduction of this distortion is employed by the current injection system. The current injection is performed by the combination of current injection network and current injection device. The purpose of these two units is to shape the input currents and to reduce the input current THD values. Predrag Pejovic (2007) has discussed the concept of optimal



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current injection for three phase diode bridge rectifier. Ali M Eltamaly (2012) has introduced a single phase controlled converter in the injection path to control the angle of injection current for each firing angle of a three phase controlled converter. He has also revealed optimal relation between the firing angle of three-phase and single phase controlled converter for minimum THD. Proposed system using closed loop controlled system.

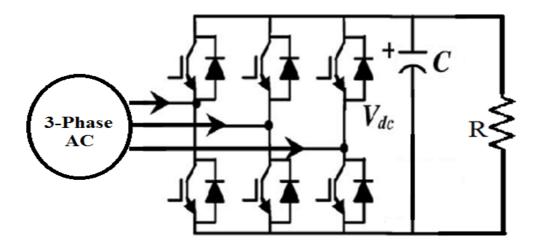


Fig-1. The circuit-diagram of Existing system

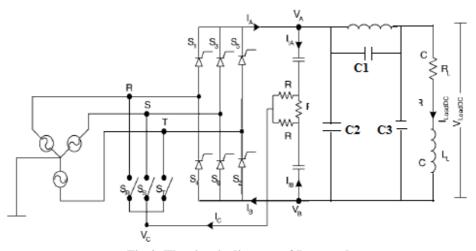


Fig-2. The circuit-diagram of Proposed system

III. SYSTEM-DESCRIPTION

"Block-Diagram of PI - controlled AC-DC controlled rectifier system is shown in Fig.3. 'Voltage at load 'is sensed and it is evaluated with the reference-voltage to get Voltage-Error (VE). This 'VE is directed to a PI-controller'. The 'yield of PI' is used to adjust the Pulse-Width (PW) of controlled Rectifier.

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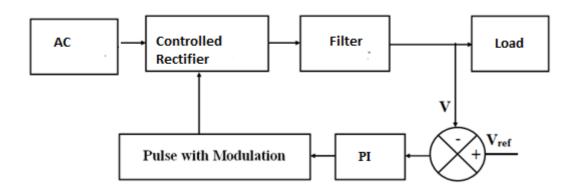


Fig.3. Block-Diagram of PI - controlled AC-DC converter system

"Block-Diagram of FOPID – controlled AC-DC controlled rectifier system is shown in Fig.4. 'Voltage at load 'is sensed and it is evaluated with the reference-voltage to get Voltage-Error (VE). This 'VE is directed to a FOPID-controller'. The 'yield of PI' is used to adjust the Pulse-Width (PW) of controlled Rectifier.

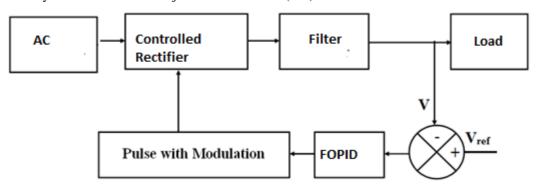


Fig.4.Block-Diagram of FOPID - controlled AC-DC converter system

IV. RESULTS AND DISCUSSION

Circuit diagram of controlled rectifier with C filter is delineated in Fig.5. Input voltage & current are delineated in Fig.6. & the values are 230V and 10A respectively. Current THD is delineated in Fig.7&its value is 42.01%. Voltage across RL load is delineated in Fig.8 &its value is 370V. Voltage Ripple is delineated in Fig.9 &its value is 362.4V. Current Through load is delineated in Fig.10 &its value is 37A. Output power is delineated in Fig.11 &its value is 13000W.

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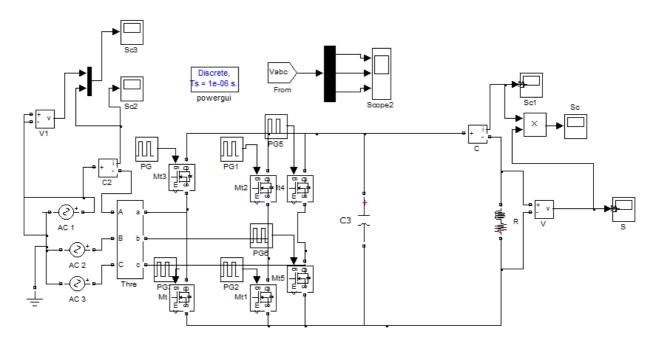


Fig.5.Circuit diagram of controlled rectifier with C filter

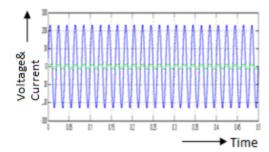


Fig.6.Input voltage & current

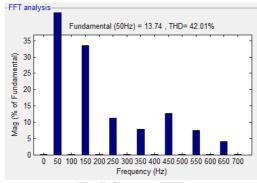


Fig.7.Current THD

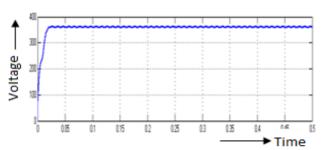


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 $Fig. 8. Voltage\ across\ RL\ load$

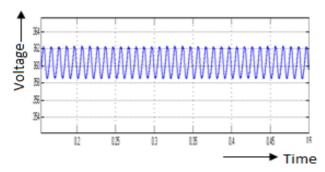


Fig.9.Voltage Ripple

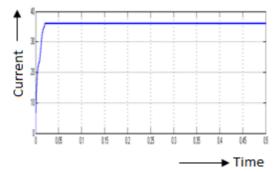


Fig.10. Current Through load

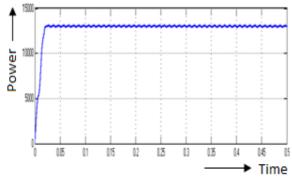


Fig.11.Output power



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Circuit diagram of controlled rectifier with Cascade filter is delineated in Fig.12. Input voltage & current are delineated in Fig.13 & the values are 260V and 30A respectively. Current THD is delineated in Fig.14 &its value is 23.93%. Voltage across RL load is delineated in Fig.15 &its value is 370V. Voltage Ripple is delineated in Fig.16 &its value is 355.5V. Current Through load is delineated in Fig.17 &its value is 37A. Output power is delineated in Fig.18 &its value is 13000W.

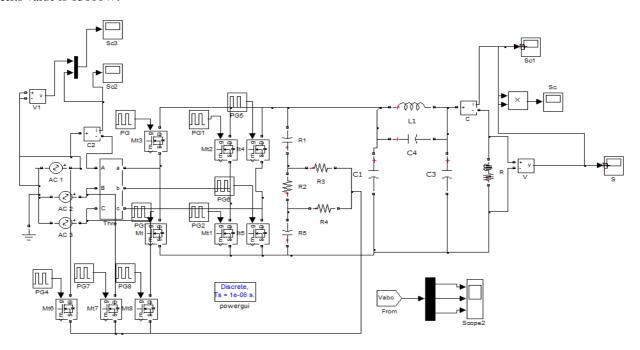


Fig.12.Circuit diagram of controlled rectifier with Cascade filter

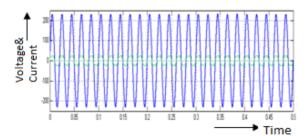


Fig.13.Input voltage and current

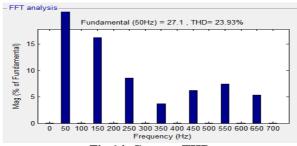


Fig.14. Current THD

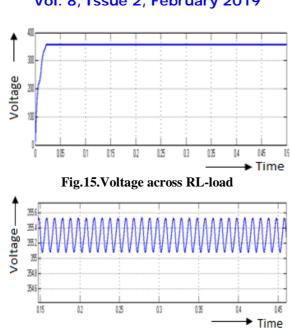


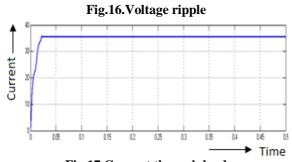
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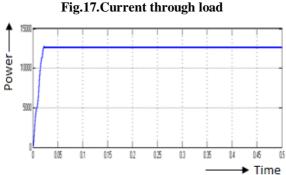


Fig.18.Output power

Comparison of voltage ripple, current THD is given in Table 1. By using cascade filter, the output voltage ripple is reduced from 3.0V to 0.6V and current THD is reduced from 42.01% to 23.93%. Hence, Controlled rectifier with cascade filter is superior to controlled rectifier with c-Filter.

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Table 1
Comparison of voltage ripple, current THD

Controlled rectifier	\mathbf{V}_{or}	Current THD
C-Filter	3.0V	42.01%
Cascade -Filter	0.6V	23.93%

Circuit diagram of controlled rectifier in open loop with disturbance is delineated in Fig.19. Input voltage is delineated in Fig.20. &its value is 420V. Output voltage across RL load is delineated in Fig.21 &its value is 490V. Output current through load is delineated in Fig.22 &its value is 44A. Output power is delineated in Fig.23 &its value is $2.2*10^4$ W.

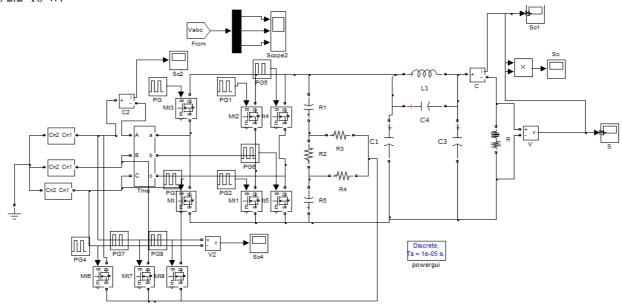


Fig.19.Circuit diagram of controlled rectifier in open loop with disturbance

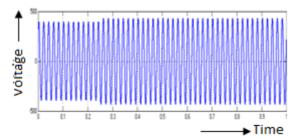


Fig.20.Input voltage



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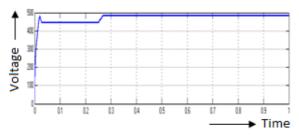


Fig.21.Output voltage across RL load

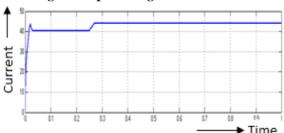


Fig.22.Output current through load

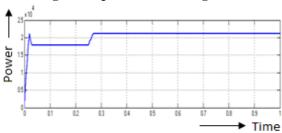


Fig.23.Output power

Circuit diagram of controlled rectifier in closed loop with PI controller is delineated in Fig.24. Input voltage is delineated in Fig.25 &its value is 420V. Output voltage across RL load is delineated in Fig.26 &its value is 460V. Output current through load is delineated in Fig.27 &its value is 46A. Output power is delineated in Fig.28 &its value is $2.2*10^4$ W.



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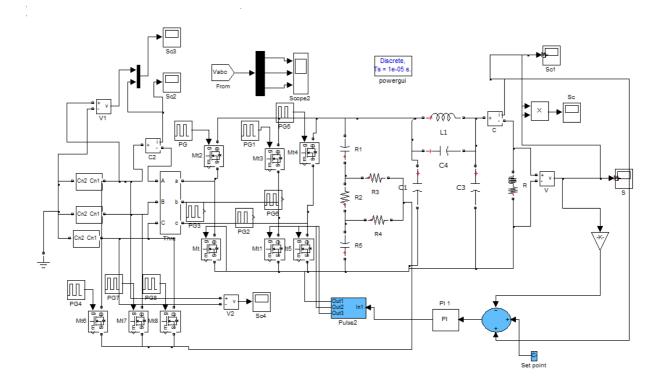


Fig.24.Circuit diagram of controlled rectifier in closed loop with PI controller

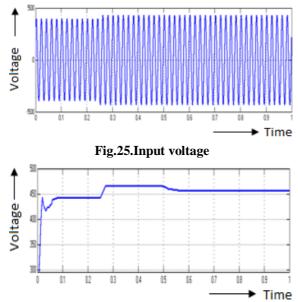


Fig.26.Output voltage across RL load

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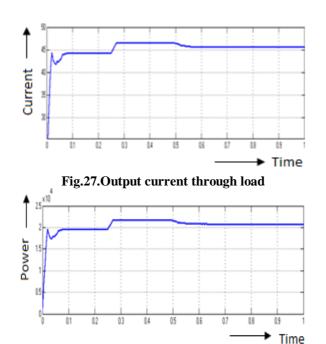


Fig.28.Output power

Circuit diagram of controlled rectifier in closed loop with FOPID controller is delineated in Fig.29. Input voltage is delineated in Fig.30 &its value is 420V. Output voltage across RL load is delineated in Fig.31 &its value is 440V. Output current through load is delineated in Fig.32 &its value is 44A. Output power is delineated in Fig.33 &its value is 2*10⁴W.

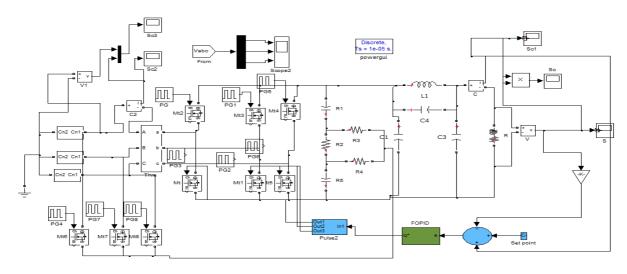


Fig.29.Circuit diagram of controlled rectifier in closed loop with FOPID controller



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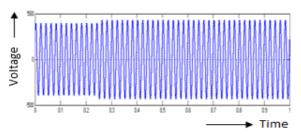


Fig.30.Input voltage

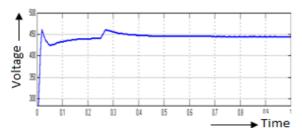


Fig.31.Output voltage across RL load

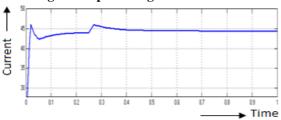


Fig.32.Output current through load

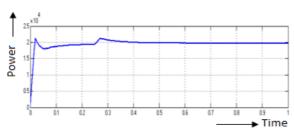


Fig.33.Output power

The comparison of Time domain parameters is given in Table-2. By using-FOPID-controller, the 'settling-time' is decreased from 0.46 to 0.35 Sec; the 'steady-state-error' is decreased from 0.6 to 0.4V; the rise-time is reduced from 0.27 Sec to 0.26 Sec; the peak-time is reduced from 0.42Sec to 0.27Seconds. Therefore, the response with controlled rectifier in closed loop with FOPID-controller is better-than-that of controlled rectifier in closed loop with PI-controller.

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Table-2 Comparison of Time Domain Parameters

Controllers	Tr (Sec)	Ts (Sec)	Tp (Sec)	Ess (V)
PI	0.27	0.46	0.42	0.6
FOPID	0.26	0.35	0.27	0.4

V.CONCLUSION

Simulation of controlled rectifier with C-Filter and Cascade filter is done using Mat lab simulink. By using cascade filter, the output voltage ripple is reduced from 3.0V to 0.6V and current THD is reduced from 42.01% to 23.93%. Both voltage ripple and current THD are reduced using cascade filter. Hence, Controlled rectifier with cascade filter is superior to controlled rectifier with c-Filter. Simulation of controlled rectifier in open loop and closed loop with PI and FOPID controller is done using Mat lab simulink. By using-FOPID-controller, the 'settling-time' is decreased from 0.46 to 0.35 Sec& the 'steady-state-error' is decreased from 0.6 to 0.4V. Settling time and steady state error is reduced using FOIPID controller. Therefore, the response with controlled rectifier in closed loop with FOPID-controller is better-than-that of controlled rectifier in closed loop with PI-controller.

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